

CHBE507 LECTURE II
MPC Revisited

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Step Response Model

• From open-loop step test

- Sampling time: Δt
- Step response coefficients: a_i
- Read the values of the unit step response

• FSR model

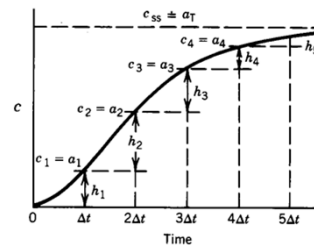
- Finite step response (FSR)
 $y_k = a_k (u_k = 1, \forall k \geq 0)$

- Using superposition principle for arbitrary input changes

$$u_k = \Delta u_0 + \Delta u_1 + \dots + \Delta u_k \text{ where } \Delta u_i = u_i - u_{i-1}$$

$$y_k = y_0 + y_k|_{\Delta u_0} + y_k|_{\Delta u_1} + \dots + y_k|_{\Delta u_{k-1}}$$

$$= y_0 + a_k \Delta u_0 + a_{k-1} \Delta u_1 + \dots + a_1 \Delta u_{k-1}$$



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Process Models

• Transfer function models

- Fixed order and structure
- Parametric: few parameters to identify
- Need very high order model for unusual behavior

• Convolution models

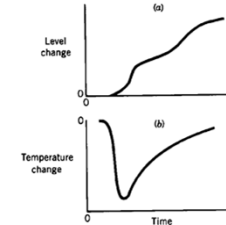
- Continuous form

$$y(t) = \int_0^t h(\tau)u(t-\tau)d\tau$$

- Discrete form

$$y(k) = \sum_{i=0}^k h(i)u(k-i)$$

Impulse response



- Many parameters, but easily obtained from the step or impulse response

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• After $t = T\Delta t$, the step response reaches steady state at least 99%

$$y_1 = y_0 + a_1 \Delta u_0$$

$$y_2 = y_0 + a_2 \Delta u_0 + a_1 \Delta u_1$$

$$y_3 = y_0 + a_3 \Delta u_0 + a_2 \Delta u_1 + a_1 \Delta u_2$$

⋮

$$y_T = y_0 + a_T \Delta u_0 + a_{T-1} \Delta u_1 + \dots + a_2 \Delta u_{T-2} + a_1 \Delta u_{T-1}$$

$$y_{T-1} = y_0 + a_T \Delta u_0 + a_T \Delta u_1 + a_{T-1} \Delta u_2 + \dots + a_2 \Delta u_{T-1} + a_1 \Delta u_T$$

$$y_{T-2} = y_0 + a_T \Delta u_0 + a_T \Delta u_1 + a_T \Delta u_2 + a_{T-1} \Delta u_3 + \dots + a_2 \Delta u_T + a_1 \Delta u_{T-1}$$

⋮

$$\Rightarrow y_n = y_0 + \sum_{i=1}^n a_i \Delta u_{n-i} \quad (a_i = a_T, \forall i \geq T) \quad \text{(FSR Model)}$$

- If there is a delay, the FSR coefficients during the delay will be zero.

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Impulse Response Model

- Impulse response coefficients**

$$h_i = a_i - a_{i-1} \quad (i = 1, 2, \dots, T)$$

$$h_0 = 0$$

$$y_n = y_0 + \sum_{i=1}^T a_i \Delta u_{n-i} = y_0 + \sum_{i=1}^T a_i (u_{n-i} - u_{n-i-1})$$

$$= y_0 + (a_1 u_{n-1} - a_1 u_{n-2}) + (a_2 u_{n-2} - a_2 u_{n-3}) + \dots + (a_n u_1 - a_n u_0) + (a_n u_0 - a_n u_{-1}) + \dots$$

$$= y_0 + a_1 u_{n-1} + (a_2 - a_1) u_{n-2} + \dots + (a_n - a_{n-1}) u_1 + (a_n - a_n) u_0 + \dots$$

$$= y_0 + (a_1 - a_0) u_{n-1} + (a_2 - a_1) u_{n-2} + \dots + (a_n - a_{n-1}) u_1$$

$$\Rightarrow y_n = y_0 + \sum_{i=1}^T h_i u_{n-i} \quad (h_i = 0, \forall i \geq T) \quad \text{(FIR Model)}$$

Matrix Form of the Predictive Model

- Horizons**

- **Model horizon:** T (number of model coefficients)
- **Control horizon:** U (number of control moves)
- **Prediction horizon:** V (number of predictions in the future)

$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \\ \vdots \\ y_V \end{bmatrix} = \begin{bmatrix} a_1 & 0 & 0 & \dots & 0 \\ a_2 & a_1 & 0 & \dots & 0 \\ \vdots & a_2 & a_1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ a_V & a_{V-1} & a_{V-2} & \dots & a_{V-U+1} \end{bmatrix} \begin{bmatrix} \Delta u_0 \\ \Delta u_1 \\ \Delta u_2 \\ \vdots \\ \Delta u_{U-1} \end{bmatrix}$$

$$\mathbf{y} = \mathbf{A} \Delta \mathbf{u}$$

- **A:** Dynamic matrix

Single-Step Prediction

- From the FIR model**

$$\hat{y}_n = y_0 + \sum_{i=1}^T h_i u_{n-i} \quad \hat{y}_{n+1} = y_0 + \sum_{i=1}^T h_i u_{n+1-i}$$

$$\Rightarrow \hat{y}_{n+1} = \hat{y}_n + \sum_{i=1}^T h_i \Delta u_{n+1-i} \quad \text{(Recursive prediction)}$$

- Corrected prediction based on the measurement**

- Assume the error between the model prediction and the measurement will present in the future with same magnitude

$$y_{n+1}^* - \hat{y}_{n+1} = y_n - \hat{y}_n \quad (y_n \text{ is the current measurement})$$

$$\Rightarrow y_{n+1}^* = \hat{y}_{n+1} + (y_n - \hat{y}_n) = y_n + \sum_{i=1}^T h_i \Delta u_{n+1-i}$$

Multi-Step Prediction

- From the single-step prediction (j -step prediction)**

$$\hat{y}_{n+j} = \hat{y}_{n+j-1} + \sum_{i=1}^T h_i \Delta u_{n+j-i} \quad (j = 1, 2, \dots, V)$$

$$y_{n+j}^* - \hat{y}_{n+j} = y_{n+j-1}^* - \hat{y}_{n+j-1} \quad (y_{n+j-1}^* \text{ is not available if } j > 1)$$

$$\Rightarrow y_{n+j}^* = y_{n+j-1}^* + \sum_{i=1}^T h_i \Delta u_{n+j-i} \quad (j = 1, 2, \dots, V)$$

- Matrix form when $V \geq U$**

$$\begin{bmatrix} y_{n+1}^* \\ y_{n+2}^* \\ y_{n+3}^* \\ \vdots \\ y_{n+V}^* \end{bmatrix} = \begin{bmatrix} a_1 & 0 & 0 & \dots & 0 \\ a_2 & a_1 & 0 & \dots & 0 \\ \vdots & a_2 & a_1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ a_V & a_{V-1} & a_{V-2} & \dots & a_{V-U+1} \end{bmatrix} \begin{bmatrix} \Delta u_n \\ \Delta u_{n+1} \\ \Delta u_{n+2} \\ \vdots \\ \Delta u_{n+U-1} \end{bmatrix} + \begin{bmatrix} y_n + P_1 \\ y_n + P_2 \\ y_n + P_3 \\ \vdots \\ y_n + P_V \end{bmatrix}$$

Dynamic Matrix, A

where

$$P_i = \sum_{j=1}^i S_j \quad (i=1, 2, \dots, V)$$

$$S_j = \sum_{i=1}^j h_i \Delta u_{n-j-i} \quad (i=1, 2, \dots, V)$$

- S_j : the incremental effect of the past (previously implemented) movements of input on the $(n+j)$ -th future output prediction (where n is current time)
- P_i : the projection which includes future prediction of y based on all previously implemented input changes.
- P_i and S_j depend only on past input changes.

- If the past information is known, then the future input changes will affect the future outputs and the future outputs can be adjusted by carefully selecting the future inputs.

Controller Design Method (DMC)

• Objective

- Minimize errors between future set points and predictions

$$\bar{\mathbf{E}} = \begin{bmatrix} r_{n+1} - y_{n+1}^* \\ r_{n+2} - y_{n+2}^* \\ \vdots \\ r_{n+V} - y_{n+V}^* \end{bmatrix} = \mathbf{r} - (\mathbf{A}\Delta\mathbf{u} + y_n \mathbf{e} + \mathbf{P}) = -\mathbf{A}\Delta\mathbf{u} + \bar{\mathbf{E}}'$$

where

$$\bar{\mathbf{E}}' = \begin{bmatrix} r_{n+1} - y_n - P_1 \\ r_{n+2} - y_n - P_2 \\ \vdots \\ r_{n+V} - y_n - P_V \end{bmatrix}$$

Closed-loop prediction error based only on current and future control action

Open-loop prediction error based only on past control action

• Solution

$$-\mathbf{A}\Delta\mathbf{u} + \bar{\mathbf{E}}' = 0 \Rightarrow \Delta\mathbf{u} = (\mathbf{A}^*)^{-1} \bar{\mathbf{E}}'$$

Some inverse of \mathbf{A}

• Currently, n is current time and y_n is measured.

$$y_{n+1}^* = y_n + \sum_{i=1}^1 h_i \Delta u_{n+1-i} = h_1 \Delta u_n + y_n + \sum_{i=2}^T h_i \Delta u_{n+1-i} = a_1 \Delta u_n + y_n + \sum_{i=2}^T h_i \Delta u_{n+1-i}$$

$$y_{n+2}^* = y_{n+1} + \sum_{i=1}^2 h_i \Delta u_{n+2-i} = (h_2 + h_1) \Delta u_n + h_2 \Delta u_{n+1} + \sum_{i=3}^T h_i \Delta u_{n+2-i} + y_n + \sum_{i=2}^T h_i \Delta u_{n+1-i}$$

$$y_{n+3}^* = y_{n+2} + \sum_{i=1}^3 h_i \Delta u_{n+3-i}$$

$$= (h_3 + h_2 + h_1) \Delta u_n + (h_2 + h_1) \Delta u_{n+1} + h_2 \Delta u_{n+2} + \sum_{i=4}^T h_i \Delta u_{n+3-i} + y_n + \sum_{i=2}^T h_i \Delta u_{n+1-i} + \sum_{i=3}^T h_i \Delta u_{n+2-i}$$

⋮

$$y_{n+V}^* = y_{n+V-1} + \sum_{i=1}^V h_i \Delta u_{n+V-i} = a_V \Delta u_n + a_{V-1} \Delta u_{n+1} + \dots + a_{V-V+1} \Delta u_{n+V-1}$$

$$+ y_n + \sum_{i=1}^{V-1} h_i \Delta u_{n+V-i} + \dots + \sum_{i=3}^T h_i \Delta u_{n+2-i} + \sum_{i=2}^T h_i \Delta u_{n+1-i}$$

$$= a_V \Delta u_n + a_{V-1} \Delta u_{n+1} + \dots + a_{V-V+1} \Delta u_{n+V-1} + y_n + \sum_{j=1}^V \sum_{i=j+1}^T h_i \Delta u_{n+i-j}$$

↑ Depend on only future

↑ Depend on only past

• If $U=V$ and \mathbf{A} is invertible,

$$\Delta\mathbf{u} = \mathbf{A}^{-1} \bar{\mathbf{E}}'$$

It gives no steady-state offset since it has integral action.

• If $U < V$ (\mathbf{A} is not invertible),

$$\Delta\mathbf{u} = (\mathbf{A}^T \mathbf{A})^{-1} \mathbf{A}^T \bar{\mathbf{E}}' = \mathbf{K}_c \bar{\mathbf{E}}'$$

\mathbf{A}^* : Left pseudoinverse of \mathbf{A}

$\mathbf{A}^* \mathbf{A} = \mathbf{I}$: identity matrix

$\mathbf{A} \mathbf{A}^*$: idempotent matrix ($\mathbf{B} \mathbf{B} = \mathbf{B}$)

• Optimization concept

$$\min(J = \bar{\mathbf{E}}^T \bar{\mathbf{E}}) = \min(-\mathbf{A}\Delta\mathbf{u} + \bar{\mathbf{E}}')^T (-\mathbf{A}\Delta\mathbf{u} + \bar{\mathbf{E}}')$$

$$\frac{\partial J}{\partial \Delta\mathbf{u}} = -2\mathbf{A}^T (-\mathbf{A}\Delta\mathbf{u} + \bar{\mathbf{E}}') = 2(\mathbf{A}^T \mathbf{A} \Delta\mathbf{u} - \mathbf{A}^T \bar{\mathbf{E}}') = 0$$

$$\Rightarrow \Delta\mathbf{u} = (\mathbf{A}^T \mathbf{A})^{-1} \mathbf{A}^T \bar{\mathbf{E}}'$$

$$\min J = (\bar{\mathbf{E}}^T \mathbf{W}_1 \bar{\mathbf{E}} + \Delta\mathbf{u}^T \mathbf{W}_2 \Delta\mathbf{u})$$

$$\frac{\partial J}{\partial \Delta\mathbf{u}} = -2\mathbf{A}^T \mathbf{W}_1 (-\mathbf{A}\Delta\mathbf{u} + \bar{\mathbf{E}}') + 2\mathbf{W}_2 \Delta\mathbf{u} = 2((\mathbf{A}^T \mathbf{W}_1 \mathbf{A} + \mathbf{W}_2) \Delta\mathbf{u} - \mathbf{A}^T \mathbf{W}_1 \bar{\mathbf{E}}') = 0$$

$$\Rightarrow \Delta\mathbf{u} = (\mathbf{A}^T \mathbf{W}_1 \mathbf{A} + \mathbf{W}_2)^{-1} \mathbf{A}^T \mathbf{W}_1 \bar{\mathbf{E}}'$$

- **Adjustable parameters of MPC (Tuning parameters)**

- **Weighting matrices**

- If $W_1 \gg W_2$, the most important objective is to minimize error of the process outputs and inputs will move quite freely.
- If $W_1 \ll W_2$, the most important objective is to minimize the input movements and controller cares much less the errors. (almost no control)
- Otherwise, it depends on the relative size of the weighting matrices.
 - If $W_1 > W_2$, aggressive action will be taken to reduce the error.
 - If $W_1 < W_2$, conservative action will be taken to reduce the input movements while reduce the error if the action is not too aggressive.
- The W_2 is called *input penalty* or *input move suppression factor*.
- Typically, use $W_1 = I$ and $W_2 = f^2 I$ and adjust f .
- If a different weighting for outputs or inputs is required, use diagonal matrix as the weighting matrix.

MIMO Extension

- **2x2 case**

$$\hat{\mathbf{E}} = -\mathbf{A}\Delta\mathbf{u} + \hat{\mathbf{E}}'$$

where

$$\hat{\mathbf{E}} = [\hat{\mathbf{E}}_1; \hat{\mathbf{E}}_2] \quad \Delta\mathbf{u} = [\Delta\mathbf{u}_1; \Delta\mathbf{u}_2]$$

$$\mathbf{A} = \begin{bmatrix} \mathbf{A}_{11} & \mathbf{A}_{12} \\ \mathbf{A}_{21} & \mathbf{A}_{22} \end{bmatrix}$$

- **General case**

- Extend the vectors and matrices in the same manner.
- If the MPC is formulated in a different form such as state-space model, different form of MIMO extension is more convenient.

- **Horizons**

- **Model horizon (T)**

- Select T such that $T\Delta t \geq$ (open-loop settling time)
- T is typically 20 to 70.

- **Prediction horizon (V)**

- Increasing V results in more conservative control action, a stabilizing effect, and more computational burden.
- An important tuning parameter

- **Control horizon (U)**

- Suitable first guess is to choose U so that $U\Delta t \cong t_{s0}$
- The larger the value of U is, the more computation time is required.
- Too large a value of U results in excessive control action
- Smaller value of U leads to a robust controller that is relatively insensitive to model error.

Constraints Handling

- **Formulate and solve the MPC in an optimization framework**

$$\min J = (\hat{\mathbf{E}}^T \mathbf{W}_1 \hat{\mathbf{E}} + \Delta\mathbf{u}^T \mathbf{W}_2 \Delta\mathbf{u})$$

$$\text{subject to } \mathbf{u}^L \leq \mathbf{u} \leq \mathbf{u}^U$$

$$\mathbf{y}^L \leq \mathbf{y} \leq \mathbf{y}^U$$

and other constraints

- **Solve this optimization problem in QP**

- DMC by DMCC used LP

Identification of Models

- **FSR or FIR models: use step or pulse test**
 - Assume operation at steady state
 - Make change in input Δu (or δu)
 - If Δu is too small, output change may not noticeable
 - If Δu is too large, linearity may not hold
 - Measure output at regular intervals Δt
 - The Δt should be chosen so that T is 20-70, typically 40.
 - Perform multiple experiments and average them and additional experiments for verification
 - High frequency information may not be accurate for step test.
 - Ideal pulse is hard to implement.

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Least Squares Identification

- Get the output using PRBS (Pseudo Random Binary Signal)

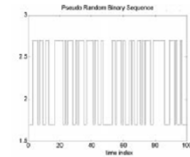
$$\mathbf{u} = [u_1 \ u_2 \ \dots \ u_M] \quad \mathbf{y} = [y_1 \ y_2 \ \dots \ y_M]$$

- Get the FIR model

$$\hat{y}_k = h_1 u_{k-1} + h_2 u_{k-2} + \dots + h_N u_{k-N}$$

- Minimize the error between measurements and output, $d_k = y_k - \hat{y}_k$

$$\begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_M \end{bmatrix} = \begin{bmatrix} u_0 & u_{-1} & \dots & u_{1-N} \\ u_1 & u_0 & \dots & u_{2-N} \\ \vdots & \vdots & \ddots & \vdots \\ u_{M-1} & u_{M-2} & \dots & u_{M-N} \end{bmatrix} \begin{bmatrix} h_1 \\ h_2 \\ \vdots \\ h_N \end{bmatrix} + \begin{bmatrix} d_1 \\ d_2 \\ \vdots \\ d_M \end{bmatrix} \quad \mathbf{d} = \mathbf{y} - \mathbf{U}\mathbf{h}$$



$$\min_{\mathbf{h}} \mathbf{d}^T \mathbf{d} = \min_{\mathbf{h}} (\mathbf{y} - \mathbf{U}\mathbf{h})^T (\mathbf{y} - \mathbf{U}\mathbf{h}) \Rightarrow \mathbf{h} = (\mathbf{U}^T \mathbf{U})^{-1} \mathbf{U}^T \mathbf{y}$$

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Discussions

- Random input testing, if appropriately designed, gives better models than the step or pulse testing does since it can equally excite low to high frequency dynamics of the process.
- If $\mathbf{U}^T \mathbf{U}$ is singular, the inverse doesn't exist and identification fails. (Need persistent excitation condition)
- When the number of coefficients is large, $\mathbf{U}^T \mathbf{U}$ can be easily singular (or nearly singular). To avoid the numerical, a regularization term is added to the cost function. (ridge regression)

$$\min_{\mathbf{h}} [(\mathbf{y} - \mathbf{U}\mathbf{h})^T (\mathbf{y} - \mathbf{U}\mathbf{h}) + \alpha \mathbf{h}^T \mathbf{h}] \Rightarrow \mathbf{h} = (\mathbf{U}^T \mathbf{U} + \alpha \mathbf{I})^{-1} \mathbf{U}^T \mathbf{y}$$

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Data Treatments

- **The data need to be processed before they are used in identification.**
- **Spike/Outlier Removal**
 - Check plots of data and remove obvious outliers (e.g., that are impossible with respect to surrounding data points). Fill in by interpolation.
 - After modeling, plot of actual vs. predicted output (using measured input and modeling equations) may suggest additional outliers. Remove and redo modeling, if necessary.
 - But don't remove data unless there is a clear justification.

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- **Bias Removal and Normalization**

- Compute the data average and subtract it to create deviation variables, i.e.,

$$\hat{y}_k = (y_k - y_{ref}) / c_y \text{ where } y_{ref} = \sum_{i=1}^M y_i / M$$

$$\hat{u}_k = (u_k - u_{ref}) / c_u \text{ where } u_{ref} = \sum_{i=1}^M u_i / M$$

- Use the given steady-state values of the variables instead to compute the deviation variables, i.e.,

$$\hat{y}_k = (y_k - y_{ss}) / c_y \text{ and } \hat{u}_k = (u_k - u_{ss}) / c_u$$

where y_{ss} and u_{ss} represent a priori given steady-state values of the process output and input respectively.

- The input/output data can be biased by the nonzero steady state and also by load disturbance effects. To remove the (time-varying) bias, differencing can be performed for the input/output data.

$$\Delta y_k = (y_k - y_{k-1}) / c_y \text{ and } \Delta u_k = (u_k - u_{k-1}) / c_u$$

⇒ Identification for Δy_k and Δu_k

- In all cases, the process data are conditioned by scaling before using in identification.

- **Prefiltering**

- If the data contain too much frequency components over an undesired range and/or if we want to obtain a model that fits well the data over a certain frequency range, data prefiltering (via digital filters) can be done.

